

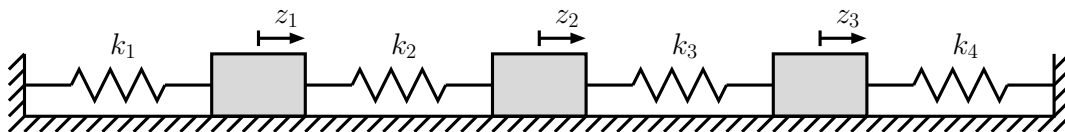
**First name:**\_\_\_\_\_**Last name:**\_\_\_\_\_**Student ID:**\_\_\_\_\_**Signature:**\_\_\_\_\_**Read before you start:**

- There are FIVE QUESTIONS.
- The examination is OPEN-BOOK.
- Besides correctness, the CLARITY of your presentation will also be graded.
- NO COLLABORATION with others.

Q1	Q2	Q3	Q4	Q5	Total

### Q1.

Consider the LTI mass spring system below, where there is no friction. The blocks are identical with unit mass, i.e.,  $m_1 = m_2 = m_3 = 1$ . The displacements of the blocks from the equilibrium positions are denoted by  $z_1, z_2, z_3 \in \mathbb{R}$  as shown in the figure. The spring constants are denoted by  $k_i > 0$  for  $i = 1, 2, 3, 4$ . For each of the following cases find (if exists) a condition on the spring constants so that the system is unobservable from the indicated output  $y$ .



(a)  $y = z_1$ .

(b)  $y = z_2$ .

**Sol'n.** The dynamics of the system read

$$\ddot{z}_1 + k_1 z_1 + k_2(z_1 - z_2) = 0 \quad (1)$$

$$\ddot{z}_2 + k_2(z_2 - z_1) + k_3(z_2 - z_3) = 0 \quad (2)$$

$$\ddot{z}_3 + k_3(z_3 - z_2) + k_4 z_3 = 0. \quad (3)$$

Recall that a system is unobservable if and only if we can find a nonzero solution which produces an identically zero output  $y(t) \equiv 0$ ; see HW4 Q1(a). This allows us to proceed as follows.

(a) Let  $y \equiv 0$ . Then  $z_1 \equiv 0 \implies \dot{z}_1 \equiv 0 \implies \ddot{z}_1 \equiv 0$ . By (1) this implies  $-k_2 z_2 \equiv 0$ . Since  $k_2 > 0$  we can write  $z_2 \equiv 0$  yielding  $\ddot{z}_2 \equiv 0$ . Now, under  $z_1 \equiv 0$ ,  $z_2 \equiv 0$ , and  $\ddot{z}_2 \equiv 0$  we have  $-k_3 z_3 \equiv 0$  thanks to (2). Then we can write  $z_3 \equiv 0$  because  $k_3$  is nonzero. To sum up, we have discovered that  $y \equiv 0$  implies  $z_1, z_2, z_3 \equiv 0$  (and, consequently,  $\dot{z}_1, \dot{z}_2, \dot{z}_3 \equiv 0$ ). In other words, only the system at rest produces  $y \equiv 0$ . Hence the system is observable from  $y = z_1$  regardless of the spring constants.

(b) Let  $y \equiv 0$ . Then we have  $z_2, \ddot{z}_2 \equiv 0$ . Rewriting the dynamics under this condition we have

$$\ddot{z}_1 + (k_1 + k_2)z_1 = 0$$

$$k_2 z_1 + k_3 z_3 = 0$$

$$\ddot{z}_3 + (k_3 + k_4)z_3 = 0.$$

Can these equations be satisfied by some nonzero solution? Yes. Let  $k_1 + k_2 = k_3 + k_4 = \omega^2$ . Then  $z_1(t) = k_3 \cos(\omega t)$  and  $z_3(t) = -k_2 \cos(\omega t)$  satisfy the above equations. Hence the system is unobservable from  $y = z_2$  if the spring constants satisfy  $k_1 + k_2 = k_3 + k_4$ .

**Q2.**

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Let  $P, Q, R \in \mathbb{R}^{n \times n}$  be symmetric positive definite matrices.

- (a) Find (if possible) a pair  $(P, Q)$  such that the following system is unstable. Explain if impossible.

$$\dot{x} = \underbrace{\begin{bmatrix} 0 & -P \\ Q & 0 \end{bmatrix}}_{A_1} x$$

- (b) Find (if possible) a triple  $(P, Q, R)$  such that the following system is stable. Explain if impossible.

$$\dot{x} = \underbrace{\begin{bmatrix} P & -R \\ R & Q \end{bmatrix}}_{A_2} x$$

**Sol'n.** (a) Let the matrix  $G \in \mathbb{R}^{2n \times 2n}$  be

$$G = \begin{bmatrix} Q & 0 \\ 0 & P \end{bmatrix}.$$

Note that  $G$  is symmetric positive definite and satisfies  $A_1^T G + G A_1 = 0$ . Let  $x(t)$  be an arbitrary solution of the system  $\dot{x} = A_1 x$ . Construct the scalar function  $V(x(t)) = x(t)^T G x(t)$ . We can write  $\dot{V} = \dot{x}^T G x + x^T G \dot{x} = x^T (A_1^T G + G A_1) x = 0$ . Therefore  $V(x(t)) = V(x(0))$  for all  $t \geq 0$ . This allows us to write  $\|x(t)\|^2 \leq \lambda_{\min}(G)^{-1} V(x(t)) = \lambda_{\min}(G)^{-1} V(x(0)) \leq \lambda_{\min}(G)^{-1} \lambda_{\max}(G) \|x(0)\|^2$ . That is,  $\|x(t)\| \leq \sqrt{\lambda_{\min}(G)^{-1} \lambda_{\max}(G)} \|x(0)\|$  for all  $t \geq 0$ . Since all its solutions are bounded the system  $\dot{x} = A_1 x$  is stable regardless of the choice  $(P, Q)$ .

- (b) Let  $H = -A_2$ . We then have

$$H^T + H = -2 \begin{bmatrix} P & 0 \\ 0 & Q \end{bmatrix}.$$

Therefore we can write  $H^T I + I H < 0$ , meaning the Lyapunov inequality is satisfied. In other words, the system  $\dot{z} = H z$  is exponentially stable. As a result, all the eigenvalues  $\lambda_i(H)$  are on the open left half-plane. Then, since  $\lambda_i(A_2) = -\lambda_i(H)$ , all the eigenvalues  $\lambda_i(A_2)$  have positive real parts. The system  $\dot{x} = A_2 x$  therefore is unstable for all  $(P, Q, R)$ .

**Q3.**

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Let  $A \in \mathbb{R}^{n \times n}$  and  $B \in \mathbb{R}^{n \times m}$ . Consider the discrete-time system  $x(k+1) = Ax(k) + Bu(k)$ . Suppose this system is stabilizable. Hence we can find a feedback gain  $K_1 \in \mathbb{R}^{m \times n}$  such that under the feedback law  $u(k) = -K_1x(k)$  the closed-loop system is exponentially stable. Does this imply that we can find a feedback gain  $K_2 \in \mathbb{R}^{m \times n}$  such that the closed-loop system under  $u(k) = -K_2x(k-1)$  is exponentially stable? Explain.

**Sol'n.** No. Here is a counterexample. Consider the first order system  $x(k+1) = 4x(k) + 1u(k)$  where  $A = 4$  and  $B = 1$ . The system clearly is controllable (hence stabilizable). Under  $u(k) = -K_2x(k-1)$  we obtain the closed-loop system  $x(k+1) = 4x(k) - K_2x(k-1)$ . Letting  $\eta(k) = [x(k-1) \ x(k)]^T$  yields the state space representation

$$\eta^+ = \underbrace{\begin{bmatrix} 0 & 1 \\ -K_2 & 4 \end{bmatrix}}_{A_{\text{cl}}} \eta.$$

For exponential stability both eigenvalues  $\lambda_1$  and  $\lambda_2$  of  $A_{\text{cl}}$  should be placed inside the unit circle by a proper choice of  $K_2$ . Note that the characteristic polynomial of  $A_{\text{cl}}$  reads  $d(s) = s^2 - 4s + K_2 = s^2 - (\lambda_1 + \lambda_2)s + \lambda_1\lambda_2$ , which tells us that  $\lambda_1 + \lambda_2 = 4$  regardless of  $K_2$ . Now, clearly,  $\lambda_1 + \lambda_2 = 4$  means at least one eigenvalue will always be outside the unit circle no matter how we choose the gain  $K_2$ . Hence this system cannot be stabilized by delayed state feedback.

**Q4.**

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Consider the following system

$$\begin{aligned}\dot{x} &= \underbrace{\begin{bmatrix} -1 & 1 & 0 \\ -2 & 0 & 2 \\ 1 & 1 & -2 \end{bmatrix}}_A x + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_B u \\ y &= \underbrace{[1 \quad -1 \quad 0]}_C x.\end{aligned}$$

(a) Is this system detectable? Explain.

(b) Is this system BIBO stable? Explain.

**Sol'n.** (a) No. Note that  $v = [1 \ 1 \ 1]^T$  is the eigenvector for the eigenvalue at the origin, i.e.,  $Av = 0$ . Moreover,  $Cv = 0$ . That is,  $A$  has an eigenvector in null  $C$  whose eigenvalue satisfies  $\text{Re}(\lambda) \geq 0$ . By the eigenvector test therefore the system is undetectable.

(b) Yes. The transfer function of this system reads

$$H(s) = C(sI - A)^{-1}B = \frac{-2}{(s+1)(s+2)}.$$

Since both poles  $\lambda_1 = -1$  and  $\lambda_2 = -2$  satisfy  $\text{Re}(\lambda_i) < 0$ , the system is BIBO stable.

**Q5.**

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Consider the following system

$$\begin{aligned} x^+ &= \underbrace{\begin{bmatrix} -1 & 1 & 0 \\ -2 & 0 & 2 \\ 1 & 1 & -2 \end{bmatrix}}_A x + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_B u \\ y &= \underbrace{[1 \quad -1 \quad 0]}_C x. \end{aligned}$$

Design the observer gain  $L \in \mathbb{R}^{3 \times 1}$  and the feedback gain  $K \in \mathbb{R}^{1 \times 3}$  such that under the observer-based feedback law  $u = -K\hat{x}$  (where  $\hat{x}$  is the state estimate generated by your observer) the closed-loop system is finite-time stable. That is, there exists some finite integer  $N$  such that all the solutions satisfy  $x(k) = 0$  for  $k \geq N$  regardless of the initial conditions  $x(0)$  and  $\hat{x}(0)$ . Write also your observer dynamics.

**Sol'n.** The observer dynamics read  $\hat{x}^+ = A\hat{x} - BK\hat{x} + L(y - C\hat{x})$ . Hence the overall closed-loop dynamics can be written as

$$\begin{bmatrix} x \\ \hat{x} \end{bmatrix}^+ = \begin{bmatrix} A & -BK \\ LC & A - BK - LC \end{bmatrix} \begin{bmatrix} x \\ \hat{x} \end{bmatrix}.$$

Rewriting the above equation in terms of  $x$  and the error  $e = \hat{x} - x$  produces

$$\begin{bmatrix} x \\ e \end{bmatrix}^+ = \underbrace{\begin{bmatrix} A - BK & -BK \\ 0 & A - LC \end{bmatrix}}_{\Phi} \begin{bmatrix} x \\ e \end{bmatrix}.$$

Finite-time stability can be achieved by making  $\Phi$  nilpotent. Since  $\Phi$  is block triangular, this is possible by making the individual blocks  $A - BK$  and  $A - LC$  on the diagonal nilpotent. In other words, the gains  $K$  and  $L$  should be chosen such that all the eigenvalues of  $A - BK$  and  $A - LC$  are at the origin. Below then is the answer.

$$K = [2.5 \quad 0.5 \quad -3], \quad L = \begin{bmatrix} 2 \\ 5 \\ 0 \end{bmatrix} + \alpha \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}$$

where  $\alpha \in \mathbb{R}$  is arbitrary.