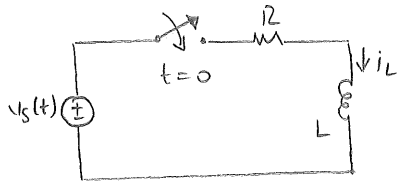


Example [sinusoidal input]Find $i_L(t)$ for $t > 0$.

$$v_s(t) = V_0 \cos(\omega t + \phi)$$

Sol'n $LDi_L = v_L = v_s - Ri_L \Rightarrow Di_L + \frac{R}{L}i_L = \frac{1}{L}v_s$

Hence $Di_L + \frac{R}{L}i_L = \frac{V_0}{L} \cos(\omega t + \phi)$ [Diff. eqn.] (Note: natural freq. $s = -\frac{R}{L}$)
 $i_L(0) = 0$ [Init. cond.]

Then homog. sol'n : $i_h(t) = Ke^{-\frac{R}{L}t}$

part. sol'n : $i_p(t) = A \cos \omega t + B \sin \omega t$ (i.e. a sinusoidal function whose freq. ω same as that of the input.)

To be determined : A, B, K (use the diff. eqn. to find A, B ; use the initial cond. constraint to find K)

$$Di_p + \frac{R}{L}i_p = \frac{V_0}{L} \cos(\omega t + \phi) \quad \{ \cos(\alpha + \beta) = \cos \alpha \cos \beta - \sin \alpha \sin \beta \}$$

$$\Rightarrow \frac{d}{dt} \{ A \cos \omega t + B \sin \omega t \} + \frac{R}{L} \{ A \cos \omega t + B \sin \omega t \} = \frac{V_0}{L} \{ \cos \phi \cos \omega t - \sin \phi \sin \omega t \}$$

$$\Rightarrow -A\omega \sin \omega t + B\omega \cos \omega t + \frac{AR}{L} \cos \omega t + \frac{BR}{L} \sin \omega t = \frac{V_0 \cos \phi}{L} \cos \omega t - \frac{V_0 \sin \phi}{L} \sin \omega t$$

$$\Rightarrow \left\{ B\omega + \frac{AR}{L} - \frac{V_0 \cos \phi}{L} \right\} \cos \omega t + \left\{ -A\omega + \frac{BR}{L} + \frac{V_0 \sin \phi}{L} \right\} \sin \omega t = 0$$

$$\underbrace{\hspace{10em}}_{=0}$$

$$\underbrace{\hspace{10em}}_{=0}$$

$$\Rightarrow \begin{bmatrix} \frac{R}{L} & \omega \\ -\omega & \frac{R}{L} \end{bmatrix} \begin{bmatrix} A \\ B \end{bmatrix} = \begin{bmatrix} \frac{V_0}{L} \cos \phi \\ -\frac{V_0}{L} \sin \phi \end{bmatrix} \Rightarrow \begin{bmatrix} A \\ B \end{bmatrix} = \frac{V_0/L}{\frac{R^2}{L^2} + \omega^2} \begin{bmatrix} \frac{R}{L} & -\omega \\ \omega & \frac{R}{L} \end{bmatrix} \begin{bmatrix} \cos \phi \\ -\sin \phi \end{bmatrix}$$

$$\Rightarrow i_L(t) = i_h(t) + i_p(t) = K e^{-\frac{R}{L}t} + A \cos \omega t + B \sin \omega t$$

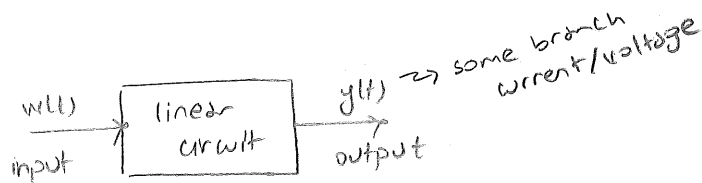
$$i_L(0) = 0 \Rightarrow K = -A$$

$$\text{Hence } i_L(t) = -A e^{-\frac{R}{L}t} + A \cos \omega t + B \sin \omega t \quad (t > 0)$$

$$\text{Let } R = 3\Omega, L = 2H, \omega = 2\text{ rad/sec}, \phi = 0^\circ, V_0 = 10V$$

$$\Rightarrow A = \frac{6}{5} \text{ and } B = \frac{8}{5} \Rightarrow i_L(t) = \underbrace{-\frac{6}{5} e^{-\frac{3}{2}t}}_{\text{transient part}} + \underbrace{\frac{6}{5} \cos 2t + \frac{8}{5} \sin 2t}_{(\text{sinusoidal}) \text{ steady state part}} \text{ Amps}$$

Superposition in linear dynamic circuits



Note that

$$y(t) = y(t, x, w(\cdot))$$

initial condition (vector) x input signal (vector) w

$$\text{init. cond. } x = [v_{c1}(0^-) \dots v_{cm}(0^-), i_{L1}(0^-) \dots i_{Ln}(0^-)]^T$$

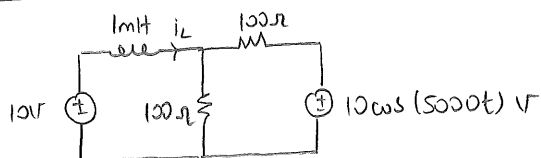
$$\text{input } w(t) = [v_{s1}(t) \dots v_{sp}(t), i_{s1}(t) \dots i_{sq}(t)]^T$$

Remark for a first-order circuit $x = v_c(0^-)$ or $x = i_L(0^-)$.

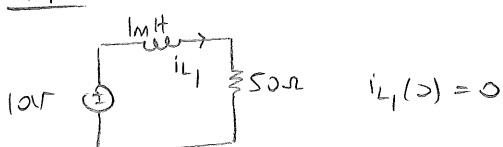
We have:

- 1) $y(t, \alpha_1 x_1 + \alpha_2 x_2, 0) = \alpha_1 y(t, x_1, 0) + \alpha_2 y(t, x_2, 0)$ superposition w.r.t. init. cond.
- 2) $y(t, 0, \beta_1 w_1 + \beta_2 w_2) = \beta_1 y(t, 0, w_1) + \beta_2 y(t, 0, w_2)$ superposition w.r.t. inputs
- 3) $y(t, \alpha x, \beta w) = \alpha y(t, x, 0) + \beta y(t, 0, w)$ superposition w.r.t. init. cond. & input

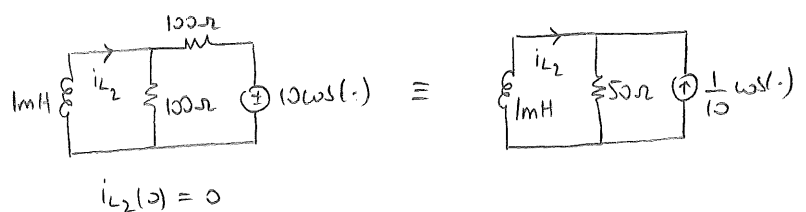
Exercise Find the zero-state response i_L



Step 1 Kill AC input to find i_{L1}



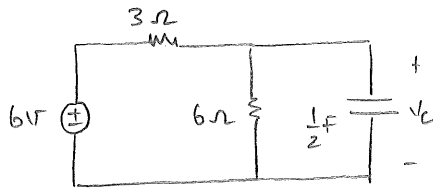
Step 2 Kill DC input to find i_{L2}



$$i_{L2}(0) = 0$$

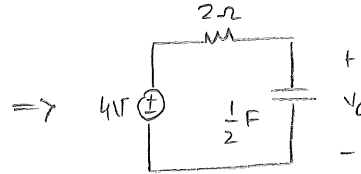
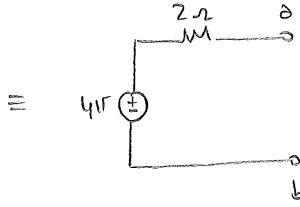
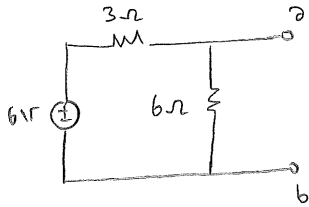
Step 3 Superpose

$$i_L(t) = i_{L1}(t) + i_{L2}(t)$$

Complete response [nonzero input + nonzero init. cond.]Example

$$v_c(0) = 2V$$

$$v_c(t) = ? \text{ for } t \geq 0$$

Sol'n

$$v_c(0) = 2V$$

$$KVL: 0 = -4 + 2i_c + v_c = -4 + 2\left\{\frac{1}{2} \frac{dv_c}{dt}\right\} + v_c \Rightarrow \frac{dv_c}{dt} + v_c = 4 \quad (\text{diff. eqn.})$$

$$\begin{aligned} \text{hom sol'n: } v_h(t) &= Ke^{-t} \\ \text{part. sol'n: } v_p(t) &= 4 \end{aligned} \quad \left. \vphantom{\begin{aligned} \text{hom sol'n: } v_h(t) &= Ke^{-t} \\ \text{part. sol'n: } v_p(t) &= 4 \end{aligned}} \right\} v_c(t) = 4 + Ke^{-t} \Rightarrow \boxed{v_c(t) = 4 - 2e^{-t} V} \quad (v_c(0) = 2V)$$

Another approach (superposition)

Step 1 Find zero-state resp. $v_{zs}(t)$

$$\frac{dv_c}{dt} + v_c = 4 \quad \text{with } v_c(0) = 0 \Rightarrow v_{zs}(t) = 4 - 4e^{-t} V$$

Step 2 Find zero-input resp. $v_{zi}(t)$

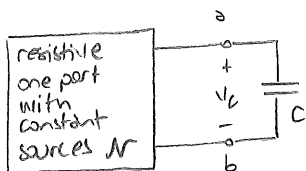
$$\frac{dv_c}{dt} + v_c = 0 \quad \text{with } v_c(0) = 2V \Rightarrow v_{zi}(t) = 2e^{-t} V$$

Step 3 Superpose

$$v_c(t) = v_{zs}(t) + v_{zi}(t)$$

$$= 4 - 2e^{-t} V$$

(as expected)

Yet another approach

$$v_c(0) = V_0$$

What do we know?

$$\rightarrow \text{form of diff eqn. } \frac{dv_c}{dt} + \frac{1}{\tau} v_c = \text{constant}$$

$$\rightarrow \text{form of sol'n } v_c(t) = A + Be^{-t/\tau} \quad (1)$$

How to find A, B, τ ?

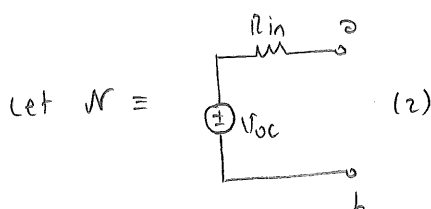
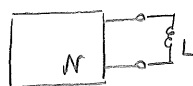
$$(1) \Rightarrow v_c(\infty) = A$$

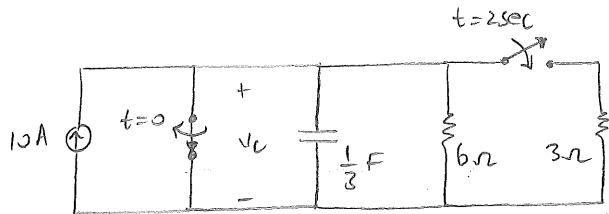
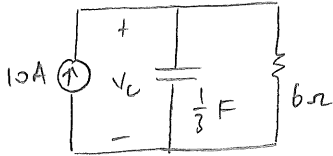
$$(2) \Rightarrow v_c(\infty) = V_{oc}$$

$$\boxed{A = V_{oc}}$$

$$(2) \Rightarrow \boxed{\tau = R_{in}C}$$

$$(1) \Rightarrow A + B = v_c(0) \Rightarrow B = v_c(0) - v_c(\infty) \Rightarrow \boxed{B = V_0 - V_{oc}}$$

Exercise Work out the dual case

ExampleFind $v_c(t)$ for $t \geq 0$.Sol'n $0 \leq t < 2$ 

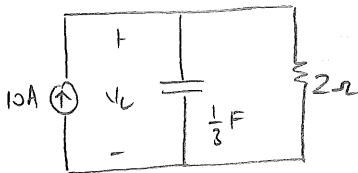
$$v_c(0) = 0$$

$$KCL \Rightarrow \frac{1}{3} \frac{dv_c}{dt} + \frac{v_c}{6} = 10 \Rightarrow \frac{dv_c}{dt} + \frac{1}{2} v_c = 30$$

$$\left. \begin{aligned} v_h(t) &= K e^{-t/2} \\ v_p(t) &= 60 \text{ V} \end{aligned} \right\} v_c(t) = 60 + K e^{-t/2} \quad \left. \begin{aligned} & \\ & \end{aligned} \right\} v_c(0) = 0$$

$$= 60 - 60 e^{-t/2} \text{ V}$$

$$\text{at } t=2 \quad v_c(2) = 60 \left(1 - \frac{1}{e}\right) =: V_2 \approx 38 \text{ V}$$

 $t \geq 2$ 

$$v_c(2) = V_2$$

$$\frac{dv_c}{dt} + \frac{3}{2} v_c = 30$$

$$\left. \begin{aligned} v_h(t) &= K e^{-\frac{3}{2}(t-2)} \\ v_p(t) &= 20 \text{ V} \end{aligned} \right\} v_c(t) = 20 + K e^{-\frac{3}{2}(t-2)} \quad \left. \begin{aligned} & \\ & \end{aligned} \right\} v_c(2) = V_2$$

$$= 20 + [V_2 - 20] e^{-\frac{3}{2}(t-2)} \text{ V}$$

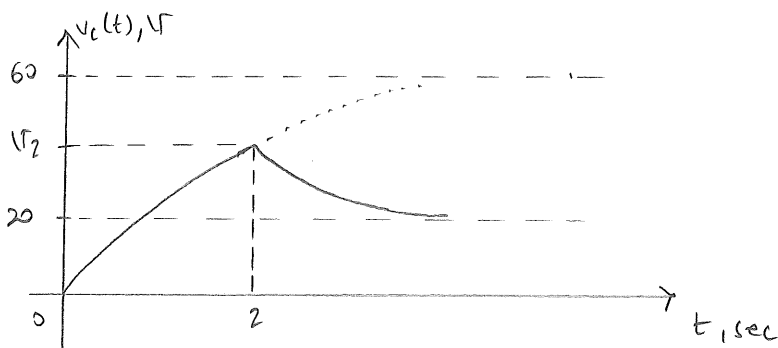
or

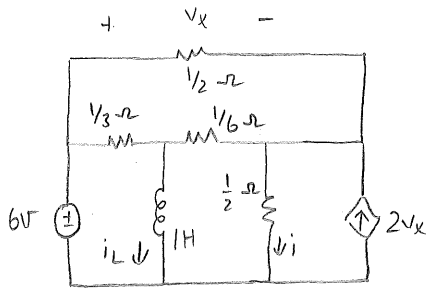
$$\left. \begin{aligned} v_h(t) &= \tilde{K} e^{-\frac{3}{2}t} \\ v_p(t) &= 20 \text{ V} \end{aligned} \right\} v_c(t) = 20 + \tilde{K} e^{-\frac{3}{2}t} \quad \left. \begin{aligned} & \\ & \end{aligned} \right\} v_c(2) = V_2$$

$$= 20 + [(V_2 - 20)e^3] e^{-\frac{3}{2}t} \text{ V}$$

Hence

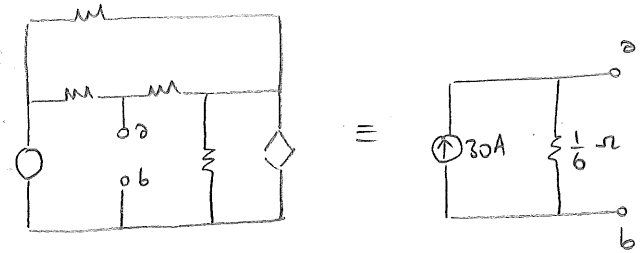
$$v_c(t) = \begin{cases} 60 [1 - e^{-t/2}] \text{ V} & \text{for } 0 \leq t < 2 \\ 20 + [V_2 - 20] e^{-\frac{3}{2}(t-2)} \text{ V} & \text{for } t \geq 2 \end{cases}$$



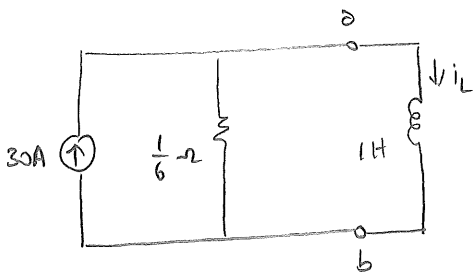
Example

$$i_L(0) = 6A, \quad i(t) = ? \text{ for } t \geq 0$$

Step 1 Find the Norton / Thevenin equiv. circuit seen by the inductor



Step 2 Using the equiv. circuit find $i_L(t)$



$$\text{KCL: } -30 + i_L + 6L \frac{di_L}{dt} = 0$$

$$\Rightarrow 6 \frac{di_L}{dt} + i_L = 30 \Rightarrow i_L(t) = 30 + K e^{-t/6}$$

$$i_L(0) = 6A \Rightarrow K = -24 \Rightarrow i_L(t) = 30 - 24e^{-t/6} A$$

$$\Rightarrow v_L(t) = 4e^{-t/6} V$$

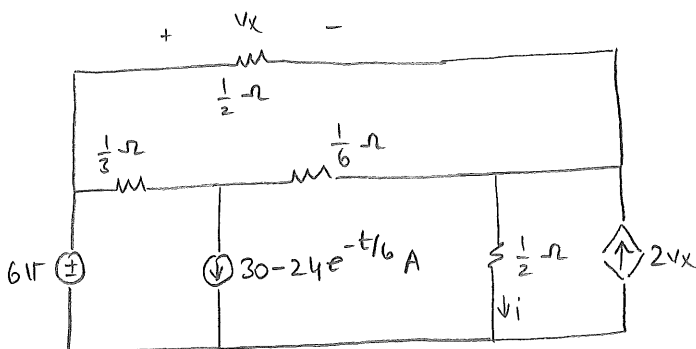
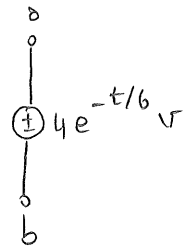
Step 3 Substitute



with



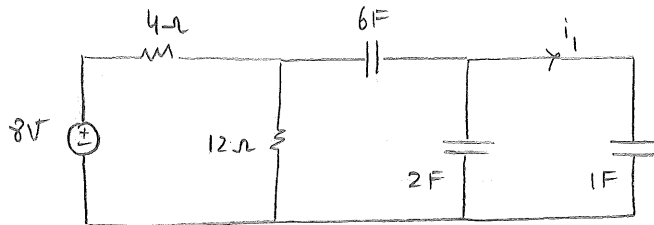
or



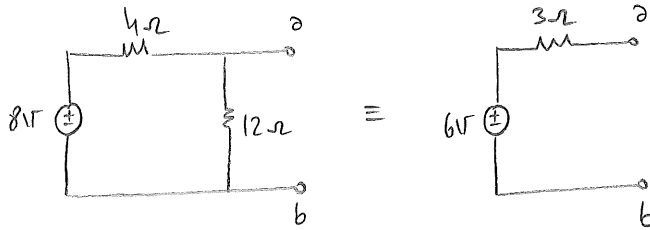
Solve for i

$$\text{Answer: } i(t) = 4 + 4e^{-t/6} A$$

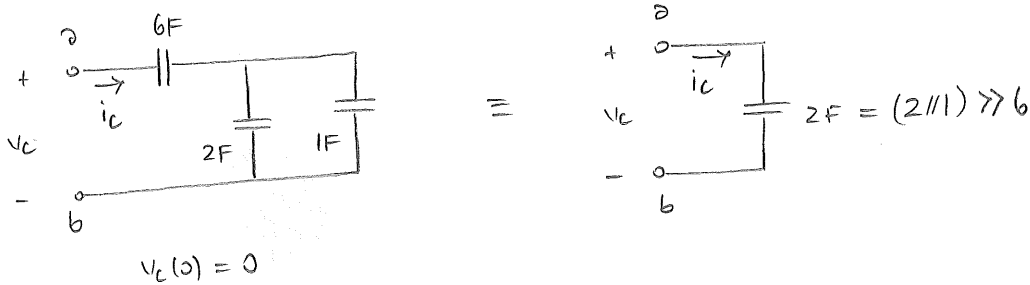
for $t \geq 0$

Example

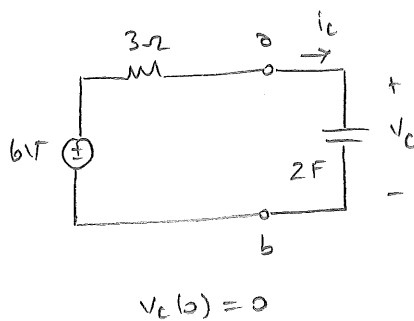
Capacitors initially uncharged.

Find $i_1(t)$ for $t \geq 0$.Sol'n

and



Hence



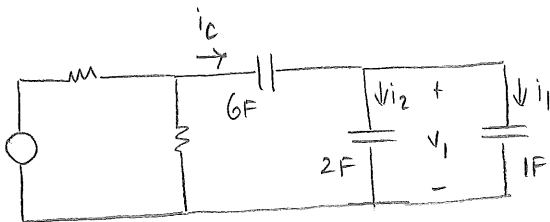
$$3(20V_c) + V_c = 6 \Rightarrow 0V_c + \frac{1}{6} V_c = 1$$

$$\Rightarrow V_c(t) = 6 + 14e^{-t/6} \quad \left. \begin{array}{l} \\ \end{array} \right\} V_c(0) = 0$$

$$= 6 - 6e^{-t/6}$$

$$\Rightarrow i_c(t) = 20V_c = 2e^{-t/6} \text{ A}$$

Return to the original configuration



$$\left. \begin{array}{l} i_1 = 10V_1 \\ i_2 = 20V_1 \end{array} \right\} \Rightarrow i_2 = 2i_1$$

$$i_c = i_1 + i_2 = 3i_1 \Rightarrow i_1 = \frac{1}{3} i_c$$

$$\Rightarrow i_1(t) = \frac{2}{3} e^{-t/6} \text{ A} \quad \text{for } t \geq 0$$