

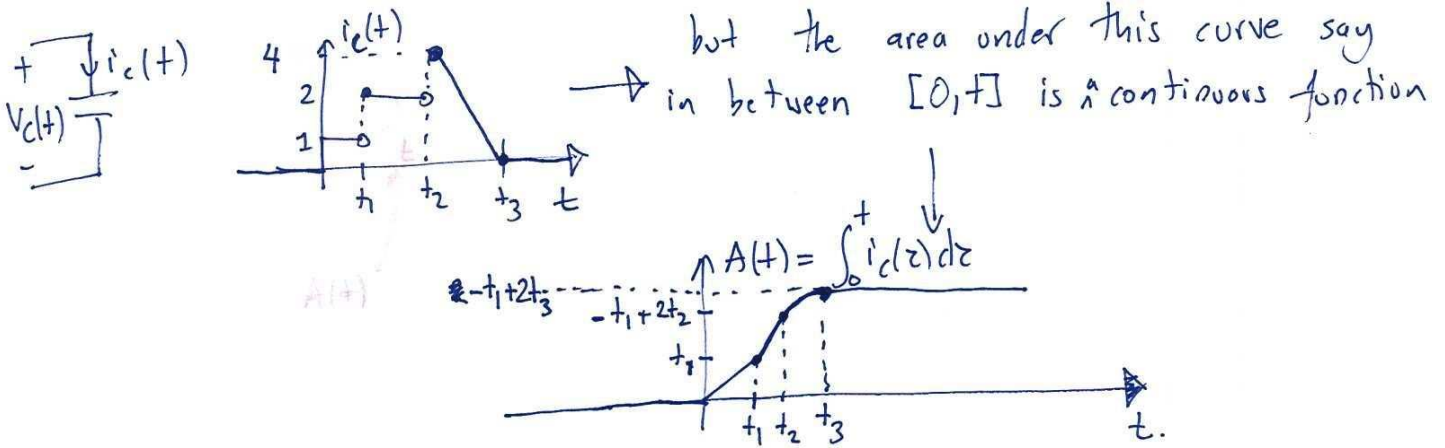
Switching Circuits - Continuity of $V_{cap}(t)$ and $I_L(t)$.

(1)

Previously, we have discussed that unless there is an impulsive source; the $V_c(t)$ and $I_L(t)$ of the circuit is a continuous function of time. Remember that, this is due to the integral operator in

$$V_c(t) = V_c(t_0^-) + \frac{1}{C} \int_{t_0^-}^t i_c(z) dz. \quad \left(\text{or } I_L(t) = I_L(t_0^-) + \frac{1}{L} \int_{t_0^-}^t V_L(z) dz \right).$$

It is clear that $i_c(z)$ can be discontinuous, as shown below



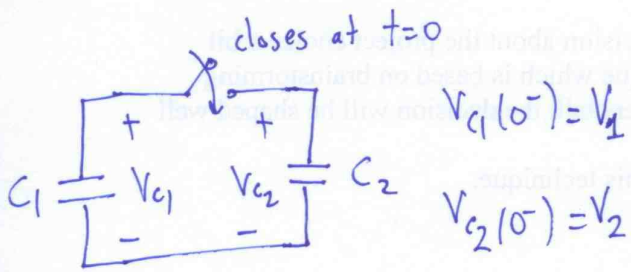
It should be clear that integrator output is "smoother" than its input! Furthermore it should be checked that $\frac{d}{dt} A(t) = i_c(t)$;

which is the Fundamental Theorem of Calculus.

So, $V_c(t)$ and $I_L(t)$ are formed by integrating a function; therefore if the integrand does not contain any impulses; we have the continuity of $V_c(t)$ and $I_L(t)$; i.e. $V_c(t_x^+) = V_c(t_x^-)$ and $I_L(t_x^+) = I_L(t_x^-)$

Discontinuity of $V_c(t)$ and $I_L(t)$ due to switching: (2)

In the following circuits, we have no sources but dynamic elements and switches. (Note that there are no resistors!)



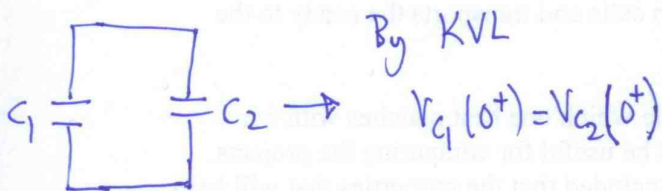
When switch is open $i_{c1} = i_{c2} = 0$;

then $(V_c(t) = V_c(t_0) + \frac{1}{C} \int_{t_0}^t i_c(z) dz)$

no change in cap. voltages. \rightarrow

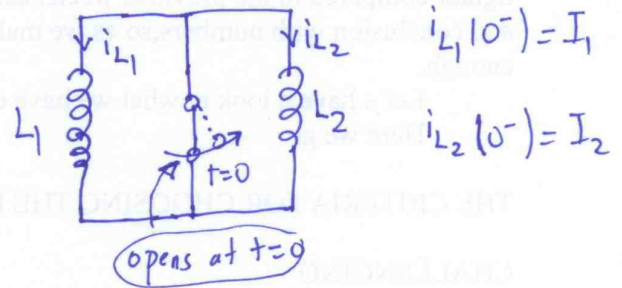
\rightarrow initial cond. are preserved.

Just after switching, i.e. $t=0^+$.



So, unless $V_{c1}(0^-) = V_{c2}(0^-)$; we have jump at the voltages after switching.

Q: What is the common voltage after switching?



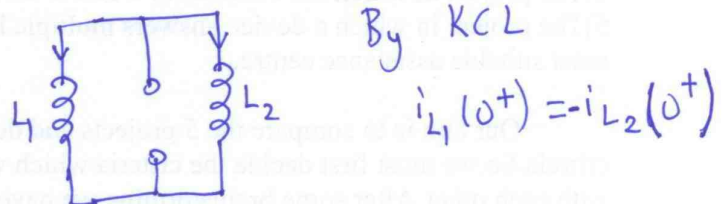
When switch is closed, $V_{L1} = V_{L2} = 0 \rightarrow$

\rightarrow No change in $i_{L1} = i_{L2} \rightarrow$

initial conditions are preserved.



At $t=0^+$



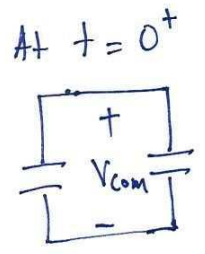
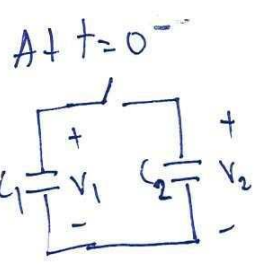
So unless $I_{L1}(0^-) = I_{L2}(0^-) \rightarrow$

\rightarrow we have a jump in inductor currents.

Q: What is the common current after switching?

Answer:

Method 1 Conservation of Charge



$$\phi(0^-) = C_1 V_1 + C_2 V_2$$

$$\phi(0^+) = C_1 V_{com} + C_2 V_{com}$$

Total charge in two caps

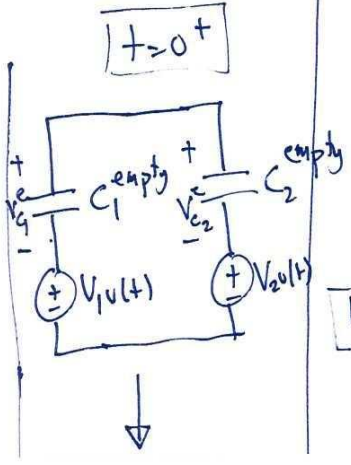
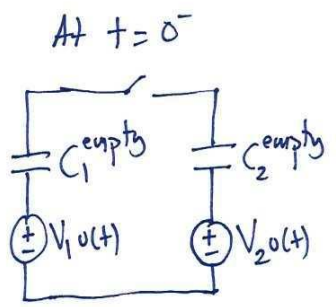
Then $\phi(0^-) = \phi(0^+)$

$$C_1 V_1 + C_2 V_2 = (C_1 + C_2) V_{com}$$

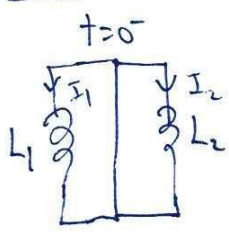
$$V_{com} = \frac{C_1 V_1 + C_2 V_2}{C_1 + C_2}$$

$$V_{C_1}(0^+) = V_{C_2}(0^+)$$

Method 2 Initial Cond. Models.

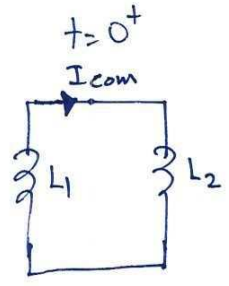


Method 1 Conservation of flux



$$\phi(0^-) = -L_1 I_1 + L_2 I_2$$

Tot. flux



$$\phi(0^+) = (L_1 + L_2) I_{com}$$

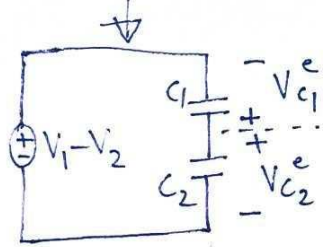
$$I_{com} = \frac{-L_1 I_1 + L_2 I_2}{L_1 + L_2}$$

Note: I_{com} is in the same direction with I_2 and it is in the opposite direction with I_1 . This is the reason of "-" sign on RHS of $\phi(0^-)$.

Comment: +/- issues related to conservation principles is too tricky and may not be required if you know initial cond. models.

Method 2 Initial Cond. Models

⋮



$$V_{C1}^{emp} = -(V_1 - V_2) \frac{1/C_1}{1/C_1 + 1/C_2} = (V_2 - V_1) \frac{C_2}{C_1 + C_2}$$

$$V_{C2}^{emp} = (V_1 - V_2) \frac{1/C_2}{1/C_1 + 1/C_2} = (V_1 - V_2) \frac{C_1}{C_1 + C_2}$$

These are voltages of the empty cap.'s; we want the voltages of actual cap.'s of the circuit.

$$V_{C1}(0^+) = V_{C1}^{emp}(0^+) + V_1 = \frac{C_1 V_1 + C_2 V_2}{C_1 + C_2}$$

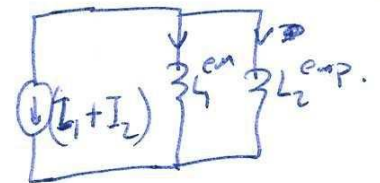
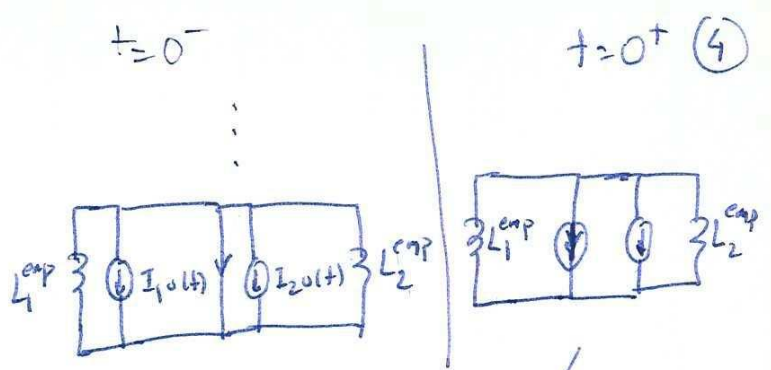
$$V_{C2}(0^+) = V_{C2}^{emp}(0^+) + V_2 = \frac{C_1 V_1 + C_2 V_2}{C_1 + C_2}$$

So,

$$V_{C1}(0^+) = V_{C2}(0^+) = \frac{C_1 V_1 + C_2 V_2}{C_1 + C_2}$$

Same as before

Let's try to use the other initial cond. model



$$I_{L1}^{emp}(0^+) = (I_1 + I_2) \cdot \frac{L_2}{L_1 + L_2}$$

$$I_{L2}^{emp}(0^+) = -(I_1 + I_2) \frac{L_1}{L_1 + L_2}$$

Actual Inductor currents:

$$I_{L1}(0^+) = I_{L1}^{emp}(0^+) + I_1 = -\frac{(L_1 I_1 + L_2 I_2)}{L_1 + L_2}$$

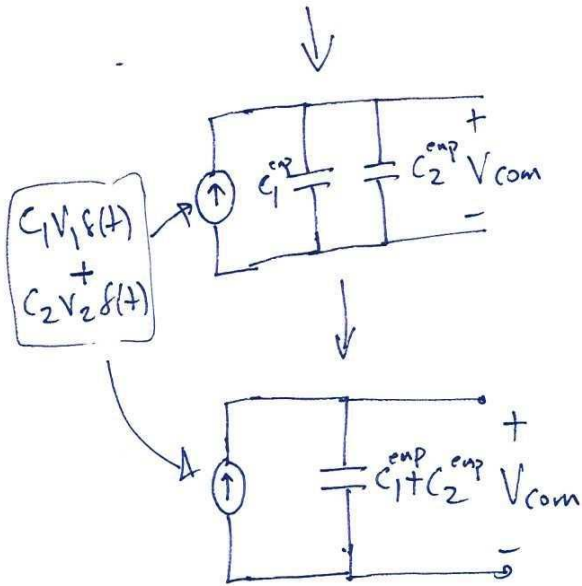
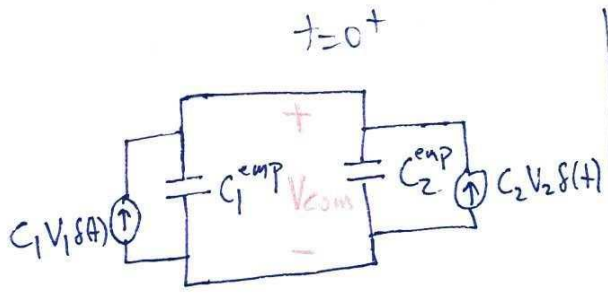
$$I_{L2}(0^+) = I_{L2}^{emp}(0^+) + I_2 = \frac{-L_1 I_1 + L_2 I_2}{L_1 + L_2}$$

$$I_{com} = I_{L2}(0^+) = -I_{L2}(0^+)$$

as before.

Let's use the other initial cond. model

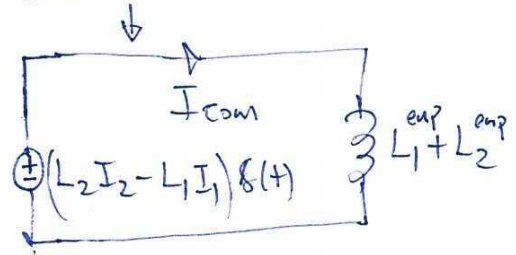
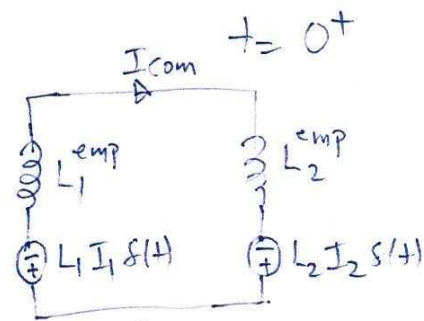




$$V_{com}(0^+) = V_{com}(0^-) + \frac{1}{C_1 + C_2} \int_{0^-}^{0^+} (C_1 V_1 + C_2 V_2) \delta(t') dt'$$

$$= \frac{1}{C_1 + C_2} (C_1 V_1 + C_2 V_2)$$

Same as before



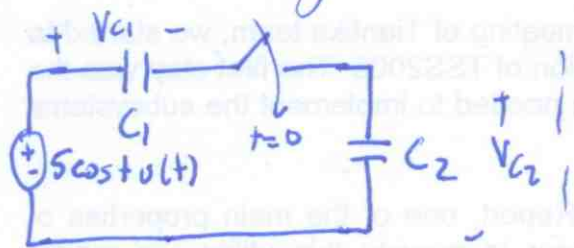
$$I_{com}(0^+) = \frac{1}{L_1 + L_2} \int_{0^-}^{0^+} V_{L_1 + L_2}(z) dz$$

$$= \frac{1}{L_1 + L_2} (L_2 I_2 - L_1 I_1)$$

(5)

The situation can be further interesting, when we have a source in the system

Ex

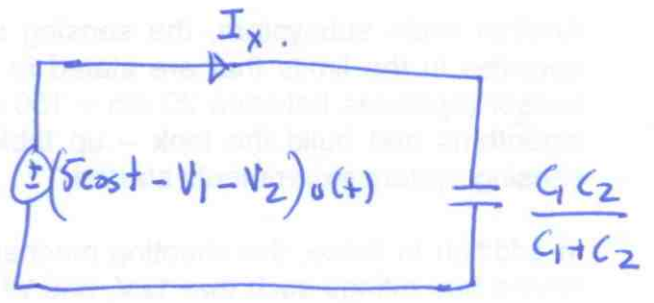
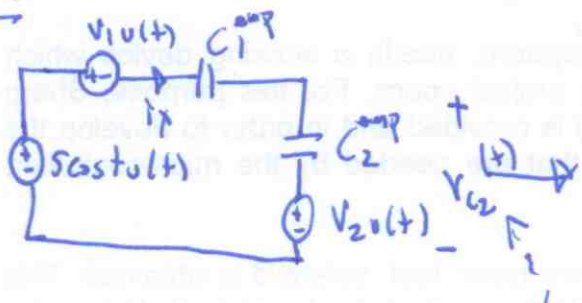


Find $V_{C2}(t)$ for $t \geq 0$.

$$V_{C1}(0^-) = V_1$$

$$V_{C2}(0^-) = V_2$$

At $t = 0^+$



$$I_x = \frac{C_1 C_2}{C_1 + C_2} \frac{d}{dt} \{ 5 \cos t - V_1 - V_2 \} u(t)$$

$$= \frac{C_1 C_2}{C_1 + C_2} (-5 \sin t u(t) + (5 - V_1 - V_2) \delta(t))$$

$$V_{C2}(t) = V_2 + \frac{1}{C_2} \int_{0^-}^t i_x(t') dt'$$

$$= V_2 + \frac{C_1}{C_1 + C_2} (5 \cos t - 5) + \frac{C_1}{C_1 + C_2} (5 - V_1 - V_2)$$

$t \geq 0$

$$= V_2 + \frac{C_1}{C_1 + C_2} 5 \cos t = \frac{C_1}{C_1 + C_2} (V_1 + V_2)$$